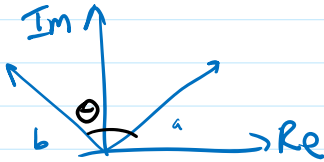


## Complex Numbers

$$z = a + jb$$

• Magnitude:  $|z| = \sqrt{a^2 + b^2}$

• Phase:  $\theta = \angle z = \text{atan2}(b, a)$   
 $-\pi \leq \text{atan2}(b, a) \leq \pi$



Polar form:  $z = |z| e^{j\theta}$

Derivation:  $a = r \cos \theta$ ,  $b = r \sin \theta$  (geometry)

$$f(a, b) \rightarrow f(r, \theta): z = a + jb = r \cos \theta + jr \sin \theta = r(\cos \theta + j \sin \theta)$$

Euler's Formula  
 $= r e^{j\theta} = |z| e^{j\theta}$

Taylor Series derivation:

$$e^x = 1 + x + \frac{x^2}{2!} + \frac{x^3}{3!} + \dots$$

So  $e^{j\theta} = 1 + j\theta + \frac{(j\theta)^2}{2!} + \frac{j\theta^3}{3!} + \dots$

$$= 1 - \frac{\theta^2}{2!} + \frac{\theta^4}{4!} - \dots + j\left(\theta - \frac{\theta^3}{3!} + \frac{\theta^5}{5!} - \dots\right)$$

$$= \sin \theta + j \cos \theta$$

Example:  $\sqrt{j} = (j)^{\frac{1}{2}} = (e^{j\frac{\pi}{2}})^{\frac{1}{2}} = e^{j\frac{\pi}{4}}$

Roots of  $z^n = 1$

• Convert 1 using Euler's Formula:

$$1 = 1 e^{0i} = e^{2\pi i} \dots = e^{2ki}$$

$$(z^n)^{\frac{1}{n}} = (1 e^{2ki})^{\frac{1}{n}}$$

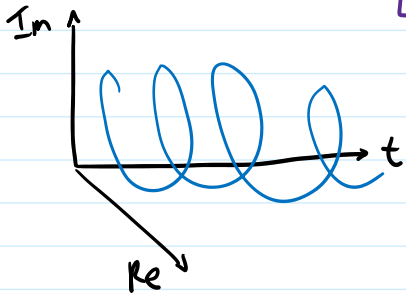
$$z = e^{2\frac{k}{n}i} \text{ for all } k \text{ such that } \frac{k}{n} < 2\pi$$

## Phasors

$$e^{j\omega t} = \cos(\omega t) + j \sin(\omega t)$$

$$e^{j\omega t} = \cos(\omega t) + j\sin(\omega t)$$

$\omega$  angular frequency (rad/s)



Scaling:  $re^{j\phi} \cdot e^{j\omega t}$       Rotating:  $1 \cdot e^{j\phi} \cdot e^{j\omega t}$

## Linear Combinations of $e^{j\omega t}$

• Sinusoids:

$$\rightarrow \cos(\omega t) = \frac{e^{j\omega t} + e^{-j\omega t}}{2}$$

$$\rightarrow \sin(\omega t) = \frac{e^{j\omega t} - e^{-j\omega t}}{2j} =$$

• Phasor sinusoids:

$$\begin{aligned} \rightarrow r\cos(\omega t + \phi) &= \frac{1}{2}(re^{j\omega t + \phi} + re^{j\omega t + \phi}) \\ &= re^{j\phi} \cdot \frac{1}{2}(e^{j\omega t} + e^{-j\omega t}) \end{aligned}$$

• Multi-frequency combinations:

$$\rightarrow r\cos(\omega_1 t) + r\cos(\omega_2 t) = \frac{1}{2}(e^{j\omega_1 t} + e^{-j\omega_1 t}) + \frac{1}{2}(e^{j\omega_2 t} + e^{-j\omega_2 t})$$

linear combination of  $e^{j\omega t}$ 's

↳ Allows us to stay in Phasor Domain

$$\tilde{V}_{in} e^{j\omega t} \rightarrow \boxed{\text{Circuit}} \rightarrow \lambda \tilde{V}_{in} e^{j\omega t}$$

Facts in Phasor Domain:

- $Z_R = R$
  - $Z_L = j\omega L$
  - $Z_C = \frac{1}{j\omega C}$
- } imaginary impedance: makes current go out of phase w/ voltage.

⇓  
Power through an imaginary impedance is returned

-  $\epsilon_C = j\omega C$  goes out of phase w/ voltage.

Power through an imaginary impedance is returned

Example: real + imaginary impedance



•  $V_o = \lambda V_{in}$  by eigenfunction

•  $V_o = V_{in} \frac{\frac{1}{j\omega C}}{j\omega C + R}$  by voltage divider

$$\frac{V_o}{V_{in}} = \frac{1 - j\omega RC}{1 + (j\omega RC)^2}$$

$$= \frac{1}{1 + (\omega RC)^2} + j \left( \frac{-\omega RC}{1 + (\omega RC)^2} \right)$$

Transfer Function  $H(j\omega)$  ←

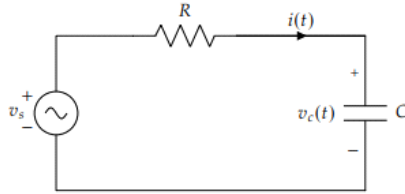
## 2 Phasor Analysis

Any sinusoidal time-varying function  $x(t)$ , representing a voltage or a current, can be expressed in the form

$$x(t) = \Re[Xe^{j\omega t}], \quad (1)$$

where  $X$  is a time-independent function called the phasor counterpart of  $x(t)$ . Thus,  $x(t)$  is defined in the time domain, while its counterpart  $X$  is defined in the phasor domain.

The phasor analysis method consists of five steps. Consider the RC circuit below.



The voltage source is given by

$$v_s(t) = 12 \sin\left(\omega t - \frac{\pi}{4}\right), \quad (2)$$

with  $\omega = 1 \times 10^3 \frac{\text{rad}}{\text{s}}$ ,  $R = \sqrt{3} \text{ k}\Omega$ , and  $C = 1 \mu\text{F}$ .

Our goal is to obtain a solution for  $i(t)$  with the sinusoidal voltage source  $v_s(t)$ .

### a) Step 1: Adopt cosine references

All voltages and currents with known sinusoidal functions should be expressed in the standard cosine format. Convert  $v_s(t)$  into a cosine and write down its phasor representation  $\tilde{V}_s$ .

$$\begin{aligned} \sin \theta &= \cos\left(\theta - \frac{\pi}{2}\right) \\ 12 \sin\left(\omega t - \frac{\pi}{4}\right) &= 12 \cos\left(\omega t - \frac{3\pi}{4}\right) \\ \text{Phasor: } \tilde{V}_s &= V_0 e^{j\phi} = 12 e^{-\frac{3\pi}{4}j} \\ &\quad \leftarrow \text{time invariant} \end{aligned}$$

b) Step 2: Transform circuits to phasor domain

The voltage source is represented by its phasor  $\tilde{V}_s$ . The current  $i(t)$  is related to its phasor counterpart  $\tilde{I}$  by

$$i(t) = \Re[\tilde{I}e^{j\omega t}]$$

What are the phasor representations of R and C?

Resistor:  $I = \frac{V}{R} \Rightarrow V_R(t) = \cos(\omega t + \phi) = V_0 e^{j\phi} \cdot e^{j\omega t}$

$$I_R(t) = \frac{V_0}{R} \cos(\omega t + \phi) = \frac{V_0}{R} e^{j\phi} e^{j\omega t}$$

Capacitor:  $I = \frac{dq}{dt} = C \frac{dV}{dt} = \frac{V_0}{j\omega C} \cos(\omega t + \phi) = \frac{V_0}{j\omega C} e^{j\phi} e^{j\omega t}$

c) Step 3: Cast KCL and/or KVL equations in phasor domain

Use Kirchhoff's laws to write down a loop equation that relates all phasors in Step 2.

$$\tilde{I}_R = \tilde{I}_C$$



$$V_s(t) - V_R(t) - V_C(t) = 0$$

$$V_s(t) - R \cdot I_R(t) - \frac{1}{j\omega C} I_C(t) = 0$$

$$\tilde{V}_s - \tilde{I}_s \left( R + \frac{1}{j\omega C} \right) = 0$$

d) Step 4: Solve for unknown variables

Solve the equation you derived in Step 3 for  $\tilde{I}$  and  $\tilde{V}_c$ . What are the polar forms of  $\tilde{I}$  ( $Ae^{j\theta}$ , where  $A$  is a positive real number) and  $\tilde{V}_c$ ?

$$\tilde{V}_s = \tilde{I}_s (R + \frac{1}{j\omega C})$$

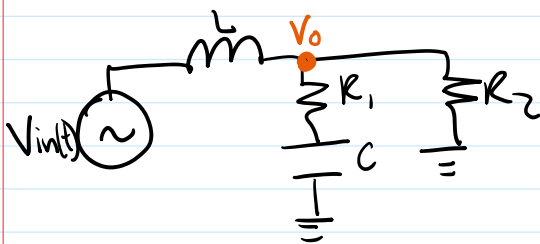
$$\tilde{I}_s = \frac{\tilde{V}_s}{R + \frac{1}{j\omega C}} \cdot \frac{j\omega C}{j\omega C} = \frac{\tilde{V}_s \cdot j\omega C}{Rj\omega C + 1}$$

$$\tilde{V}_c = \tilde{I}_s \cdot \frac{1}{j\omega C}$$

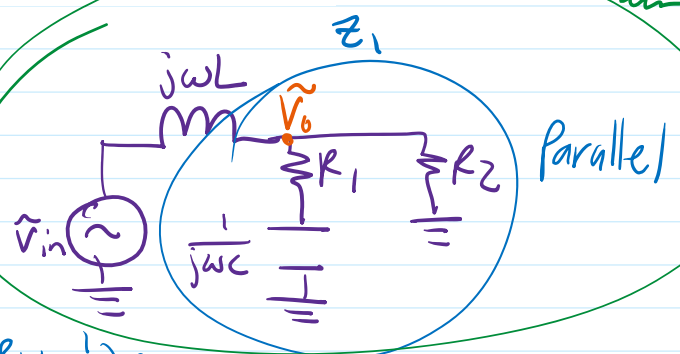
e) Step 5: Transform solutions back to time domain

To return to time domain, we apply the fundamental relation between a sinusoidal function and its phasor counterpart. What is  $i(t)$  and  $v_c(t)$ ? What is the phase difference between  $i(t)$  and  $v_c(t)$ ?

## Transfer Function Examples



Phasor Domain



$$Z_1 = \frac{(R_1 + \frac{1}{j\omega C}) R_2}{(R_1 + \frac{1}{j\omega C}) + R_2}$$

$$Z_2 = \frac{Z_1}{Z_1 + j\omega L} = H(j\omega)$$

Magnitude =  $|H(j\omega)| \rightarrow$  numpy. absolute

$$= \sqrt{R_1^2 + \frac{\omega^2 L^2}{\dots}}$$

$$= \frac{\sqrt{1 + \frac{\omega^2}{\omega_z^2}}}{\sqrt{1 + \frac{\omega^2}{\omega_p^2}}}$$

Phase angle  $\angle H(j\omega) \rightarrow \text{numpy.angle}$

$$\frac{e^{j\theta_1}}{e^{j\theta_2}} = e^{j\theta_1 - \theta_2} \text{ so } \angle H(j\omega) = \angle \left( \frac{1 + \frac{j\omega}{\omega_z}}{1 + \frac{j\omega}{\omega_p}} \right)$$

$$= \angle \left( 1 + \frac{j\omega}{\omega_z} \right) - \angle \left( 1 + \frac{j\omega}{\omega_p} \right)$$

$$= \text{atan2} \left( 1 + \frac{\omega}{\omega_z} \right) - \text{atan2} \left( 1 + \frac{\omega}{\omega_p} \right)$$

Output:  $V_{out}(t) = |H(j\omega)| \cdot V_{in}(t + \angle H(j\omega))$